# 12-Output, 76V, Serial-Interfaced VFD Tube Driver 

## General Description

The MAX6920 is a 12-output, 76V, vacuum fluorescent display (VFD) tube driver that interfaces a multiplexed VFD tube to a VFD controller such as the MAX6850-MAX6853 or to a microcontroller. The MAX6920 is also ideal for driving either static VFD tubes or telecom relays.
Data is inputted using an industry-standard 4-wire serial interface (CLOCK, DATA, LOAD, BLANK) for compatibility with both industry-standard drivers and Maxim's VFD controllers.
For easy display control, the active-high BLANK input forces all driver outputs low, turning the display off, and automatically puts the MAX6920 into shutdown mode. Display intensity may also be controlled by pulse-width modulating the BLANK input.
The MAX6920 has a serial interface data output pin, DOUT, allowing any number of devices to be cascaded on the same serial interface.
The MAX6920 is available in a 20-pin SO package. Maxim also offers VFD drivers with either 20 (MAX6921/MAX6931) or 32 outputs (MAX6922 and MAX6932).

|  | Applications |
| :--- | :--- |
| White Goods | Industrial Weighing |
| Gaming Machines | Security |
| Automotive | Telecom |
| Avionics |  |

Pin Configuration


Features

- 5 MHz Industry-Standard 4-Wire Serial Interface
- 3V to 5.5V Logic Supply Range
- 8V to 76V Grid/Anode Supply Range
- Push-Pull CMOS High-Voltage Outputs
- Outputs can Source 40mA, Sink 4mA Continuously
- Outputs can Source 75mA Repetitive Pulses
- Outputs can be Paralleled for Higher Current Drive
- Any Output can be Used as a Grid or an Anode Driver
- Blank Input Simplifies PWM Intensity Control
- Small 20-Pin SO Package
- $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ Temperature Range

Ordering Information

| PART | TEMP RANGE | PIN-PACKAGE |
| :---: | :---: | :--- |
| MAX6920AWP | $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | 20 Wide SO |

Typical Operating Circuit


## 12-Output, 76V, Serial-Interfaced VFD Tube Driver

## ABSOLUTE MAXIMUM RATINGS




OUT_Sink Current .............................................................. 15 mA
CLK, DIN, LOAD, BLANK, DOUT Current ....................... $\pm 10 \mathrm{~mA}$
Continuous Power Dissipation Junction Temperature
$+150^{\circ} \mathrm{C}$
Storage Temperature Range ............................. $-65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Lead Temperature (soldering, 10s)
$+300^{\circ} \mathrm{C}$
 operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## ELECTRICAL CHARACTERISTICS

(Typical Operating Circuit, $\mathrm{V}_{\mathrm{BB}}=8 \mathrm{~V}$ to $76 \mathrm{~V}, \mathrm{~V}_{\mathrm{CC}}=3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=\mathrm{T}_{\mathrm{MIN}}$ to $\mathrm{T}_{\mathrm{MAX}}$, unless otherwise noted.) (Note 1)

| PARAMETER | SYMBOL | CONDITIONS |  | MIN TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Logic Supply Voltage | VCC |  |  | 3 | 5.5 | V |
| Tube Supply Voltage | $V_{\text {BB }}$ |  |  | 8 | 76 | V |
| Logic Supply Operating Current | Icc | All outputs OUT_ low, CLK = idle | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 72 | 170 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |  | 200 |  |
|  |  | All outputs OUT_ high, CLK = idle | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 350 | 650 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |  | 700 |  |
| Tube Supply Operating Current | ${ }^{\prime} \mathrm{BB}$ | All outputs OUT_ low | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 1 | 2 | mA |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |  | 4.2 |  |
|  |  | All outputs OUT_ high | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 0.53 | 0.85 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |  | 0.9 |  |
| High-Voltage OUT_ | VH | $\begin{aligned} & \mathrm{V}_{\mathrm{BB}} \geq 15 \mathrm{~V}, \\ & \text { IOUT }=-25 \mathrm{~mA} \end{aligned}$ | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | VBB-1. |  | V |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | VBB-2 |  |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | VBB - 2.5 |  |  |
|  |  | $\mathrm{V}_{\mathrm{BB}} \geq 15 \mathrm{~V}$, | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | VBB - 3.5 |  |  |
|  |  | lout $=-40 \mathrm{~mA}$ | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | $V_{B B}-4.0$ |  |  |
|  |  | $\begin{aligned} & 8 \mathrm{~V}<\mathrm{V}_{\mathrm{BB}}<15 \mathrm{~V}, \\ & \text { IOUT }=-25 \mathrm{~mA} \end{aligned}$ | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | VBB-1.2 |  |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | VBB - 2.5 |  |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ | VBB - 3.0 |  |  |
| Low-Voltage OUT_ | VL | $\begin{aligned} & V_{\mathrm{BB}} \geq 15 \mathrm{~V}, \\ & \text { IOUT }=1 \mathrm{~mA} \end{aligned}$ | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 0.75 | 1 | V |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |  | 1.5 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |  | 1.9 |  |
|  |  | $\begin{aligned} & 8 \mathrm{~V}<\mathrm{V}_{\mathrm{BB}}<15 \mathrm{~V}, \\ & \text { IOUT }=1 \mathrm{~mA} \end{aligned}$ | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ | 0.8 | 1.1 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ |  | 1.6 |  |
|  |  |  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |  | 2.0 |  |

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## ELECTRICAL CHARACTERISTICS (continued)

(Typical Operating Circuit, $\mathrm{V}_{\mathrm{BB}}=8 \mathrm{~V}$ to $76 \mathrm{~V}, \mathrm{~V}_{\mathrm{CC}}=3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=\mathrm{T}_{\mathrm{MIN}}$ to $\mathrm{T}_{\mathrm{MAX}}$, unless otherwise noted.) (Note 1)

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Rise Time OUT_ ( $20 \%$ to 80\%) | tR | $\mathrm{V}_{\mathrm{BB}}=60 \mathrm{~V}, \mathrm{CLL}=50 \mathrm{pF}, \mathrm{R}_{\mathrm{L}}=2.3 \mathrm{k} \Omega$ |  | 0.9 | 2 | $\mu \mathrm{s}$ |
| Fall Time OUT_ ( 80\% to 20\%) | $\mathrm{t}_{\mathrm{F}}$ | $\mathrm{V}_{\mathrm{BB}}=60 \mathrm{~V}, \mathrm{CLL}=50 \mathrm{pF}, \mathrm{R}_{\mathrm{L}}=2.3 \mathrm{k} \Omega$ |  | 0.6 | 1.5 | $\mu \mathrm{s}$ |

SERIAL INTERFACE TIMING CHARACTERISTICS

| LOAD Rising to OUT_ Falling Delay |  | (Notes 2, 3) |  |  | 0.9 | 1.8 | $\mu \mathrm{S}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| LOAD Rising to OUT_ Rising Delay |  | (Notes 2, 3) |  |  | 1.2 | 2.4 | $\mu \mathrm{S}$ |
| BLANK Rising to OUT_ Falling Delay |  | (Notes 2, 3) |  |  | 0.9 | 1.8 | $\mu \mathrm{S}$ |
| BLANK Falling to OUT_ Rising Delay |  | (Notes 2, 3) |  |  | 1.3 | 2.5 | $\mu \mathrm{S}$ |
| Input Leakage Current CLK, DIN, LOAD, BLANK | IIH, IIL |  |  |  | 0.05 | 10 | $\mu \mathrm{A}$ |
| Logic-High Input Voltage CLK, DIN, LOAD, BLANK | $\mathrm{V}_{\mathrm{IH}}$ |  |  | $\begin{aligned} & 0.8 x \\ & V_{C C} \end{aligned}$ |  |  | V |
| Logic-Low Input Voltage CLK, DIN, LOAD, BLANK | VIL |  |  |  |  | $\begin{aligned} & 0.3 x \\ & V_{C C} \end{aligned}$ | V |
| Hysteresis Voltage DIN, CLK, LOAD, BLANK | $\Delta \mathrm{V}_{\mathrm{l}}$ |  |  |  | 0.6 |  | V |
| High-Voltage DOUT | VOH | ISOURCE $=-1.0 \mathrm{~mA}$ |  | $\begin{gathered} \mathrm{V}_{\mathrm{CC}}- \\ 0.5 \end{gathered}$ |  |  | V |
| Low-Voltage DOUT | VoL | $\mathrm{ISINK}=1.0 \mathrm{~mA}$ |  |  |  | 0.5 | V |
| Rise and Fall Time DOUT |  | $\begin{aligned} & \text { CDOUT = 10pF } \\ & \text { (Note 2) } \end{aligned}$ | 3 V to 4.5 V |  | 60 | 100 | ns |
|  |  |  | 4.5 V to 5.5V |  | 30 | 80 |  |
| CLK Clock Period | tcP |  |  | 200 |  |  | ns |
| CLK Pulse-Width High | tch |  |  | 90 |  |  | ns |
| CLK Pulse-Width Low | tCL |  |  | 90 |  |  | ns |
| CLK Rise to LOAD Rise Hold | tcSH | (Note 2) |  | 100 |  |  | ns |
| DIN Setup Time | tDS |  |  | 5 |  |  | ns |
| DIN Hold Time | tD | 3 V to 4.5 V |  | 20 |  |  | ns |
|  |  | 4.5 V to 5.5V |  | 15 |  |  |  |
| DOUT Propagation Delay | tDO | CDout $=10 \mathrm{pF}$ | 3.0 V to 4.5V | 25 | 120 | 240 | ns |
|  |  |  | 4.5 V to 5.5 V | 20 | 75 | 150 |  |
| LOAD Pulse High | tcsw |  |  | 55 |  |  | ns |

Note 1: All parameters are tested at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$. Specifications over temperature are guaranteed by design.
Note 2: Guaranteed by design.
Note 3: Delay measured from control edge to when output OUT_ changes by 1 V .

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$\left(\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{BB}}=76 \mathrm{~V}\right.$, and $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$, unless otherwise noted. $)$



LOGIC SUPPLY CURRENT (Icc) vs. TEMPERATURE (OUTPUTS LOW)



OUTPUT VOLTAGE
vs. TEMPERATURE (OUTPUT LOW)


OUTPUT VOLTAGE (VBB - $\mathrm{V}_{\mathrm{H}}$ ) vs. TEMPERATURE (OUTPUT HIGH)


OUTPUT RISE AND FALL WAVEFORM


## 12-Output, 76V, Serial-Interfaced VFD Tube Driver

Pin Description

| PIN | NAME | FUNCTION |
| :---: | :---: | :--- |
| 1 | VBB | VFD Tube Supply Voltage |
| 2 | DOUT | Serial-Clock Output. Data is clocked out of the internal shift register to DOUT on CLK's rising edge. |
| $3-8,13-18$ | OUTO to <br> OUT11 | VFD Anode and Grid Drivers. OUTO to OUT11 are push-pull outputs swinging from VBB to GND. |
| 9 | BLANK | Blanking Input. High forces outputs OUTO to OUT11 low, without altering the contents of the output <br> latches. Low enables outputs OUTO to OUT11 to follow the state of the output latches. |
| 10 | GND | Ground |
| 11 | CLK | Serial-Clock Input. Data is loaded into the internal shift register on CLK's rising edge. |
| 12 | LOAD | Load Input. Data is loaded transparently from the internal shift register to the output latch while LOAD <br> is high. Data is latched into the output latch on LOAD's rising edge, and retained while LOAD is low. |
| 19 | DIN | Serial-Data Input. Data is loaded into the internal shift register on CLK's rising edge. |
| 20 | VCC | Logic Supply Voltage |



Figure 1. MAX6920 Functional Diagram

## 12-Output, 76V, Serial-Interfaced VFD Tube Driver



Figure 2. MAX6920 CMOS Output Driver Structure

## Detailed Description

The MAX6920 is a VFD tube driver comprising a 4 -wire serial interface driving 12 high-voltage Rail-to-Rail $®$ output ports. The driver is suitable for both static and multiplexed displays.
The output ports feature high current-sourcing capability to drive current into grids and anodes of static or multiplex VFDs. The ports also have active current sinking for fast discharge of capacitive display electrodes in multiplexing applications.
The 4 -wire serial interface comprises a 12 -bit shift register and a 12-bit transparent latch. The shift register is written through a clock input CLK and a data input DIN and the data propagates to a data output DOUT. The data output allows multiple drivers to be cascaded and operated together. The output latch is transparent to
the shift register outputs when LOAD is high, and latches the current state on the falling edge of LOAD.
Each driver output is a slew-rated controlled CMOS push-pull switch driving between $\mathrm{V}_{\mathrm{BB}}$ and GND. The output rise time is always slower than the output fall time to avoid shoot-through currents during output transitions. The output slew rates are slow enough to minimize EMI, yet are fast enough so as not to impact the typical $100 \mu \mathrm{~s}$ digit multiplex period and affect the display intensity.

Initial Power-Up and Operation
An internal reset circuit clears the internal registers of the MAX6920 on power-up. All outputs OUT0 to OUT11 and the interface output DOUT initialize low regardless of the initial logic levels of the CLK, DIN, BLANK, and LOAD inputs.

4-Wire Serial Interface
The MAX6920 uses a 4-wire serial interface with three inputs (DIN, CLK, LOAD) and a data output (DOUT). This interface is used to write output data to the MAX6920 (Figure 3) (Table 1). The serial interface data word length is 12 bits, D0-D11.
The functions of the four serial interface pins are:

- CLK input is the interface clock, which shifts data into the MAX6920's 12-bit shift register on its rising edge.
- LOAD input passes data from the MAX6920's 12bit shift register to the 12-bit output latch when LOAD is high (transparent latch), and latches the data on LOAD's falling edge.


Figure 3. 4-Wire Serial Interface Timing Diagram
Rail-to-Rail is a registered trademark of Nippon Motorola, Ltd.

# 12-Output, 76V, Serial-Interfaced VFD Tube Driver 

Table 1. 4-Wire Serial Interface Truth Table

| SERIAL DATA | CLOCK INPUT | SHIFT REGISTER CONTENTS |  |  |  |  |  | LOAD INPUT | LATCH CONTENTS |  |  |  |  |  | BLANKING INPUT | OUTPUT CONTENTS |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DIN | CLK | D0 | D1 | D2 | ... | Dn-1 | Dn | LOAD | D0 | D1 | D2 | ... | Dn-1 | Dn | BLANK | D0 | D1 | D2 | ... | Dn-1 | Dn |
| H | $\sim$ | H | R0 | R1 | ... | Rn-2 | Rn-1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| L | $\sim$ | L | R0 | R1 | ... | Rn-2 | Rn -1 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| X | $\sim$ | R0 | R1 | R2 | $\ldots$ | Rn-1 | Rn |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
|  |  | X | X | X |  | X | X | L | R0 | R1 | R2 | $\ldots$ | Rn-1 | Rn |  |  |  |  |  |  |  |
|  |  | PO | P1 | P2 | $\ldots$ | Pn-1 | Pn | H | PO | P1 | P2 | ... | Pn-1 | Pn | L | P0 | P1 | P2 | ... | Pn-1 | Pn |
|  |  |  |  |  |  |  |  |  | X | X | X | $\ldots$ | X | X | H | L | L | L | $\ldots$ | L | L |

$L=$ Low logic level.
$H=$ High logic level.
$X=$ Don't care.
$P=$ Present state (shift register).
$R=$ Previous state (latched).

- DIN is the interface data input, and must be stable when it is sampled on the rising edge of CLK.
- DOUT is the interface data output, which shifts data out from the MAX6920's 12-bit shift register on the falling edge of CLK. Data at DIN is propagated through the shift register and appears at DOUT (20 CLK cycles + tDo) later.
A fifth input pin, BLANK, can be taken high to force outputs OUT0 to OUT11 low, without altering the contents of the output latches. When the BLANK input is low, outputs OUT0 to OUT11 follow the state of the output latches. A common use of the BLANK input is PWM intensity control.
The BLANK input's function is independent of the operation of the serial interface. Data can be shifted into the serial interface shift register and latched regardless of the state of BLANK.


## Writing Device Registers Using the 4-Wire

 Serial InterfaceThe MAX6920 is written using the following sequence:

1) Take CLK Iow.
2) Clock 12 bits of data in order D11 first to D0 last into DIN, observing the data setup and hold times.
3) Load the 12 output latches with a falling edge on LOAD.

LOAD may be high or low during a transmission. If LOAD is high, then the data shifted into the shift register at DIN appears at the OUTO to OUT11 outputs.
CLK and DIN may be used to transmit data to other peripherals. Activity on CLK always shifts data into the MAX6920's shift register. However, the MAX6920 only updates its output latch on the rising edge of LOAD, and the last 12 bits of data are loaded. Therefore, multiple devices can share CLK and DIN as long as they have unique LOAD controls.

## Determining Driver Output Voltage Drop

The outputs are CMOS drivers, and have a resistive characteristic. The typical and maximum sink and source output resistances can be calculated from the $\mathrm{V}_{\mathrm{H}}$ and $\mathrm{V}_{\mathrm{L}}$ electrical characteristics. Use this calculated resistance to determine the output voltage drop at different output currents.

## Output Current Ratings

The continuous current source capability is 40 mA per output. Outputs may drive up to 75 mA as a repetitive peak current, subject to the on time (output high) being no longer than 1 ms , and the duty cycle being such that the output power dissipation is no more than the dissipation for the continuous case. The repetitive peak rating allows outputs to drive a higher current in multiplex grid driver applications, where only one grid is on at a time, and the multiplex time per grid is no more than 1 ms .

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Since dissipation is proportional to current squared, the maximum current that can be delivered for a given multiplex ratio is given by:

IPEAK $=(\text { grids } \times 1600)^{1 / 2} \mathrm{~mA}$
where grids is the number of grids in a multiplexed display.
This means that a duplex application (two grids) can use a repetitive peak current of 56.5 mA , a triplex application (three grids) can use a repetitive peak current of 69.2 mA , and higher multiplex ratios are limited to 75 mA .

## Paralleling Outputs

Any number of outputs within the same package may be paralleled in order to raise the current drive or reduce the output resistance. Only parallel outputs directly (by shorting outputs together) if the interface control can be guaranteed to set the outputs to the same level. Although the sink output is relatively weak (typically $750 \Omega$ ), that resistance is low enough to dissipate 530 mW when shorted to an opposite level output at a $V_{B B}$ voltage of only 20 V . A safe way to parallel outputs is to use diodes to prevent the outputs from sinking current (Figure 4). Because the outputs cannot sink current from the VFD tube, an external discharge resistor, $R$, is required. For static tubes, $R$ can be a large value such as $100 \mathrm{k} \Omega$. For multiplexed tubes, the value of the resistor can be determined by the load capacitance and timing characteristics required. Resistor RI discharges tube capacitance C to $10 \%$ of the initial voltage in $2.3 \times \mathrm{RC}$ seconds. So, for example, a $15 \mathrm{k} \Omega$ value for $R$ discharges 100pF tube grid or anode from 40 V to 4 V in $3.5 \mu \mathrm{~s}$, but draws an additional 2.7 mA from the driver when either output is high.

## Power Dissipation

Take care to ensure that the maximum package dissipation ratings for the chosen package are not exceeded. Over dissipation is unlikely to be an issue when driving static tubes, but the peak currents are usually


Figure 4. Paralleling Outputs
higher for multiplexed tubes. When using multiple driver devices, try to share the average dissipation evenly between the drivers.
Determine the power dissipation (PD) for the MAX6920 for static tube drivers with the following equation:

$$
\begin{aligned}
P_{D}= & \left(\mathrm{V}_{\mathrm{CC}} \times \mathrm{I} \mathrm{CC}\right)+\left(\mathrm{V}_{\mathrm{BB}} \times \mathrm{IBB}\right)+\left(\left(\mathrm{V}_{\mathrm{BB}}-\mathrm{V}_{\mathrm{H}}\right) \times\right. \\
& \text { IANODE } \times \mathrm{A}))
\end{aligned}
$$

where:
A = number of anodes driven (a MAX6920 can drive a maximum of 12).
IANODE $=$ maximum anode current.
$\left(\mathrm{V}_{\mathrm{BB}}-\mathrm{V}_{\mathrm{H}}\right)$ is the output voltage drop at the given maximum anode current lout.
A static tube dissipation example follows:
$V_{C C}=5 \mathrm{~V} \pm 5 \%, V_{B B}=10 \mathrm{~V}$ to $18 \mathrm{~V}, \mathrm{~A}=12$, $\mathrm{IOUT}=2 \mathrm{~mA}$
$P D=(5.25 \mathrm{~V} \times 0.7 \mathrm{~mA})+(18 \mathrm{~V} \times 0.9 \mathrm{~mA})+((2.5 \mathrm{~V} \times$ $2 \mathrm{~mA} / 25 \mathrm{~mA}) \times 2 \mathrm{~mA} \times 12)=24.7 \mathrm{~mW}$
Determine the power dissipation (PD) for the MAX6920 for multiplex tube drivers with the following equation:
$P_{D}=\left(V_{C C} \times I_{C C}\right)+\left(V_{B B} \times I_{B B}\right)+\left(\left(V_{B B}-V_{H}\right) \times I_{\text {ANODE }}\right.$

$$
\left.x A)+\left(\left(V_{B B}-V_{H}\right) \times I_{G R I D}\right)\right)
$$

where:
A = number of anodes driven
$G=$ number of grids driven
IANODE = maximum anode current
IGRID = maximum grid current
The calculation presumes all anodes are on but only one grid is on. The calculated PD is the worst case, presuming one digit is always being driven with all its anodes lit. Actual PD can be estimated by multiplying this PD figure by the actual tube drive duty cycle, taking into account interdigit blanking and any PWM intensity control.
A multiplexed tube dissipation example follows:

$$
\begin{aligned}
& \mathrm{V} C \mathrm{C}=5 \mathrm{~V} \pm 5 \%, \mathrm{VBB}=36 \mathrm{~V} \text { to } 42 \mathrm{~V}, \mathrm{~A}=6, \mathrm{G}=6, \\
& \mathrm{I}_{\mathrm{ANODE}}= 0.4 \mathrm{~mA}, \mathrm{I} \mathrm{GRID}=24 \mathrm{~mA} \\
& \mathrm{PD}=(5.25 \mathrm{~V} \times 0.7 \mathrm{~mA})+(42 \mathrm{~V} \times 0.9 \mathrm{~mA})+((2.5 \mathrm{~V} \times \\
&0.4 \mathrm{~mA} / 25 \mathrm{~mA}) \times 0.4 \mathrm{~mA} \times 6)+ \\
&((2.5 \mathrm{~V} \times 24 \mathrm{~mA} / 25 \mathrm{~mA}) \times 24 \mathrm{~mA})=99 \mathrm{~mW}
\end{aligned}
$$

Thus, for a 20-pin wide SO package ( $T_{J A}=1 / 0.01=$ $+100^{\circ} \mathrm{C} / \mathrm{W}$ from Absolute Maximum Ratings), the maximum allowed ambient temperature $T_{A}$ is given by:

```
\(T_{J}(\mathrm{MAX})=\mathrm{T}_{\mathrm{A}}+\left(\mathrm{PD} \times \mathrm{T}_{\mathrm{JA}}\right)=+150^{\circ} \mathrm{C}=\mathrm{T}_{\mathrm{A}}+(0.099 \times\)
        \(\left.+100^{\circ} \mathrm{C} / \mathrm{W}\right)\)
So \(T_{A}=+140^{\circ} \mathrm{C}\).
```


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This means that the driver can be operated in this application up to the MAX6920's $+125^{\circ} \mathrm{C}$ maximum operating temperature.

## Power-Supply Considerations

The MAX6920 operates with multiple power-supply voltages. Bypass the $V_{C C}$ and $V_{B B}$ power-supply pins to GND with a $0.1 \mu \mathrm{~F}$ capacitor close to the device. For multiplex applications, it may be necessary to add an additional $1 \mu \mathrm{~F}$ bulk electrolytic capacitor, or greater, to the $V_{B B}$ supply.

Power-Supply Sequencing
The order of the power-supply sequencing is not important. The MAX6920 will not be damaged if either VCC or $V_{B B}$ is grounded (or maintained at any other voltage below the data sheet minimum), while the other supply is maintained up to its maximum rating. However, as with any CMOS device, do not drive the MAX6920's logic inputs if the logic supply $\mathrm{V}_{\mathrm{CC}}$ is not operational because the input protection diodes clamp the signals.

Chip Information
TRANSISTOR COUNT: 2743 PROCESS: BiCMOS

Typical Application Circuit


## 12-Output, 76V, Serial-Interfaced VFD Tube Driver

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to www.maxim-ic.com/packages.)


|  | INCHES |  | MILLIMETERS |  |
| :--- | :---: | :---: | :---: | :---: |
| DIM | MIN | MAX | MIN | MAX |
| A | 0.093 | 0.104 | 2.35 | 2.65 |
| A1 | 0.004 | 0.012 | 0.10 | 0.30 |
| B | 0.014 | 0.019 | 0.35 | 0.49 |
| C | 0.009 | 0.013 | 0.23 | 0.32 |
| e | 0.050 |  | 1.27 |  |
| E | 0.291 | 0.299 | 7.40 | 7.60 |
| H | 0.394 | 0.419 | 10.00 | 10.65 |
| L | 0.016 | 0.050 | 0.40 | 1.27 |

VARIATIONS:

|  | INCHES |  | MILLIMETERS |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DIM | MIN | MAX | MIN | MAX | N | MS013 |  |
| D | 0.398 | 0.413 | 10.10 | 10.50 | 16 | AA |  |
| D | 0.447 | 0.463 | 11.35 | 11.75 | 18 | AB |  |
| D | 0.496 | 0.512 | 12.60 | 13.00 | 20 | AC |  |
| D | 0.598 | 0.614 | 15.20 | 15.60 | 24 | AD |  |
| D | 0.697 | 0.713 | 17.70 | 18.10 | 28 | AE |  |


NOTES:

1. D\&E DO NOT INCLUDE MOLD FLASH.
2. MOLD FLASH OR PROTRUSIONS NOT TO EXCEED 0.15 mm (.006").
3. LEADS TO BE COPLANAR WITHIN 0.10 mm (.004").
4. CONTROLLING DIMENSION: MILLIMETERS.
5. MEETS JEDEC MSO13.
6. $\mathrm{N}=$ NUMBER OF PINS.

